FINITE RATE OF INNOVATION WITH NON-UNIFORM SAMPLES

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ABSTRACT

In this paper, we investigate the problem of retrieving the innovation parameters (time and amplitude) of a stream of Diracs from *non-uniform samples* taken with a novel kernel (a hyperbolic secant). We devise a *non-iterative, exact* algorithm that allows perfect reconstruction of 2K innovations from as few as 2K non-uniform samples. We also investigate noise issues and compute the Cramér-Rao lower bounds for this problem. A simple total least-squares extension of the algorithm proves to be efficient in reconstructing the location of a single Dirac from noisy measurements.

Index Terms— Signal sampling, finite rate of innovation, non-uniform, hyperbolic secant function, Cramér-Rao Bounds

1. INTRODUCTION

Sampling non-bandlimited — and in particular, sparse — signals has been a hot topic in recent years (see for example the special issue [1] and the papers therein). One approach is to modelize such signals using a finite number of parameters — a.k.a innovations — per unit of time, hence their name: "finite rate of innovation" (FRI) signals [2]. Examples of such signals include streams of Diracs, piecewise polynomials and piecewise sinusoidals. Very efficient SVD-based algorithms exist when the sampling kernels are periodized sinc kernels [3], modulated Gaussians [4], Strang-Fix kernels [5, 6] and even Cauchy-like analytic kernels [7].

However, one of the constraints of the FRI algorithms to date has been that the samples of the non-bandlimited signal have to be *uniform*, in order to achieve perfect reconstruction. Our contribution in this paper is twofold:

- 1. We present a new sampling kernel, the hyperbolic secant, that allows to sample and perfectly reconstruct streams of Diracs with a noniterative algorithm.
- 2. We demonstrate the possibility to perform non-uniform sampling with the new hyperbolic secant kernel, while

still being able to perform exact reconstruction using a non-iterative algorithm.

More specifically, from a stream of K Dirac impulses with amplitudes x_k and located at t_k , k = 1, 2, ..., K, we consider the set of N arbitrary samples at $t = \tau_n$

$$y_n = \sum_{k=1}^{K} x_k \varphi(\tau_n - t_k) \tag{1}$$

where $\varphi(t)$ is the sampling kernel, and propose a novel reconstruction algorithm in the case where this kernel is the *hyperbolic secant*:

$$\varphi(t) = \frac{1}{\cosh(at)}, \quad \text{where } a > 0.$$
 (2)

Problems involving level-crossing A/D converters [8] or importance sampling [9] may benefit from these new algorithms. In general, this theory may have implications in all the domains where non-uniform sampling occurs such as for example geophysics, biomedical imaging and communications. For a review on the canonical non-uniform sampling problem and its applications, we refer to [10].

The outline of the paper is as follows. In Section 2 we develop the algorithm that enables the reconstruction from nonuniform measurements. In Section 3, we discuss the noise issues and present an algorithm to deal with it. We show simulation results that evaluate the performance of our algorithm in the presence of noise in Section 4 and finally conclude in Section 5.

2. SAMPLING SIGNALS WITH FRI

By replacing the expression of the hyperbolic secant (2) in (1) we find that

$$y_n = \sum_{k=1}^{K} \frac{x_k}{\cosh(a\tau_n - at_k)} \quad \text{for } n = 1, 2, \dots, N \quad (3)$$
$$= \frac{e^{a\tau_n} P(e^{2a\tau_n})}{Q(e^{2a\tau_n})}$$

where P(x) and Q(x) are polynomials of degree K-1 and K, respectively, defined by $P(x)=\sum_{k=0}^{K-1}p_kx^k$ and Q(x)=

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 $\sum_{k=0}^{K} q_k x^k$. Moreover, Q(x) contains the information of the Dirac locations through the factorization

$$Q(x) = \prod_{k=1}^{K} \left(1 + e^{-2at_k} x \right).$$
 (4)

By multiplying both sides of (3) by Q(x), we observe that the coefficients of P and Q satisfy the following linear system of equations:

$$\sum_{k=0}^{K} q_k \mathrm{e}^{2a\tau_n k} y_n = \sum_{k=0}^{K-1} p_k \mathrm{e}^{a\tau_n(2k+1)}$$
(5)

for n = 1, 2, ..., N. This system can be expressed in matrix/vector form as follows:

$$\underbrace{ \begin{bmatrix} y_1 & 0 & \cdots & 0 \\ 0 & y_2 & \cdots & 0 \\ 0 & 0 & \ddots & 0 \\ 0 & 0 & \cdots & y_N \end{bmatrix} }_{\mathbf{A}_N} \underbrace{ \begin{bmatrix} 1 & e^{2a\tau_1} & \cdots & e^{2Ka\tau_1} \\ 1 & e^{2a\tau_2} & \cdots & e^{2Ka\tau_2} \\ \vdots & \vdots & \ddots & \vdots \\ 1 & e^{2a\tau_N} & \cdots & e^{2Ka\tau_N} \end{bmatrix} }_{\mathbf{V}_{K+1}} \underbrace{ \begin{array}{c} q_0 \\ q_1 \\ \vdots \\ q_K \\ q_K \\ \mathbf{V}_{K+1} \\ \mathbf{q}_{K} \\ \mathbf{$$

The system of N equations $(N \ge 2K)$ can be solved for the 2K unknowns (up to a multiplicative constant). From its coefficients, we compute the roots, z_k , of Q(x), from which we retrieve the Dirac locations, according to

$$t_k = \frac{\ln(-z_k)}{2a}.\tag{7}$$

Having the locations t_k , the amplitude x_k can be obtained by minimizing the mean square difference between the parametric FRI model and the samples, which once again amounts to solving a linear system of equations. We thus conclude that perfect reconstruction of Diracs is possible with the hyperbolic secant kernel and using only $N \ge 2K$ samples. Reconstruction using samples at the critical sampling rate is exemplified in Figure 1.

3. RECONSTRUCTION SCHEME IN THE PRESENCE OF NOISE

The previous section demonstrates an ideal situation where no noise exists. However, in practice, any acquisition device introduces noise during the acquisition process. Hence we are looking for a reconstruction algorithm that can estimate the innovation parameters from the noisy samples:

$$\hat{y}_n = \sum_{k=1}^K x_k \varphi \big(\tau_n - t_k \big) + \epsilon_n \tag{8}$$



(6) Fig. 1. Retrieval of K = 2 Diracs from N = 4 nonuniform samples. Fig. 1(a) shows the two Diracs, the 4 non uniform samples, the signal ∑^K_{k=1} x_kφ(τ − t_k) where φ(t) is the secant kernel (continuous line) and the two functions x₁φ(τ − t₁) and x₂φ(τ − t₂) (dashed lines). Fig. 1(b): original and reconstructed Diracs.

where the noise may be, for instance, additive Gaussian with zero mean and variance σ^2 . To combat the noise, we propose a variation of the algorithm used to solve for t_k in order to make the scheme more robust.

Given that the τ_n are sorted in increasing order of value, we notice that the entries in the two matrices V_{K+1} and W_K increase when going from top left to bottom right and these entries will increase more rapidly for large values of *a* (recall that *a* depends on the sampling kernel as indicated in Eq. (2)). This results in a large dynamics of matrix coefficients, which is likely to impair the accuracy of the computations. To alleviate this problem, both sides of the linear system (6) are multiplied by the following weighting matrix

$$\mathbf{M} = \begin{bmatrix} e^{-Ka\tau_1} & 0 & \cdots & 0\\ 0 & e^{-Ka\tau_2} & \cdots & 0\\ 0 & 0 & \ddots & 0\\ 0 & 0 & \cdots & e^{-Ka\tau_N} \end{bmatrix}.$$
 (9)

Note that this practice does not really eradicate the problem, but improves the decomposition accuracy a bit. In other words, it provides similar decomposing accuracy for a comparatively larger a.

To further improve stability we propose the following ap-

proach. First of all, the following relation holds:

$$\Delta_N \mathbf{V}'_{K+1} \mathbf{q} = \mathbf{W}'_K \mathbf{p} \tag{10}$$

where $\mathbf{V}'_{K+1} = \mathbf{M}\mathbf{V}_{K+1}$ and $\mathbf{W}'_{K+1} = \mathbf{M}\mathbf{W}_{K+1}$.

Next, we eliminate the right hand side of Equation (10) by looking for an orthonormal $(N-K) \times N$ matrix \mathbf{U}_{N-K} such that $\mathbf{U}_{N-K}^T \mathbf{W}_K' = \mathbf{0}^{1}$. In this way we obtain the following new relation:

$$\mathbf{U}_{N-K}^{T} \boldsymbol{\Delta}_{N} \mathbf{V}_{K+1}^{\prime} \mathbf{q} = \mathbf{0}.$$
 (11)

This is the annihilation equation that the polynomial Q(x) satisfies and which is familiar to other FRI schemes. After QR decomposition of \mathbf{V}'_{K+1} into $\mathbf{Q}_{K+1}\mathbf{R}$, the above relation becomes

$$\underbrace{\mathbf{U}_{N-K}^{T}\boldsymbol{\Delta}_{N}\mathbf{Q}_{K+1}}_{\mathbf{A}}\mathbf{R}\mathbf{q}=\mathbf{0}.$$
 (12)

Notice that now both \mathbf{U}_{N-K} and \mathbf{Q}_{K+1} are orthonormal matrices and this provides the stability in the computations that we were looking for. In the presence of noise, however, (12) is not satisfied exactly. Nevertheless, a good estimation of **Rq** can be obtained by looking for a solution that minimizes $\|\mathbf{ARq}\|$ under the constrain that $\|\mathbf{Rq}\|^2 = 1$. It is known that this minimization can be solved by performing Singular Value Decomposition (SVD) of **A**.

The eigenvector corresponding to the smallest eigenvalue given by this SVD procedure represents the desired estimation of **Rq**. Thus **q** can be obtained by $\mathbf{R}^{-1}\mathbf{R}\mathbf{q}$. Finally, given **q**, we follow the steps indicated in Section 2 to reconstruct the signal. More precisely, we can get the roots z_k of Q(x) to then estimate the innovation instants t_k from (7).

4. SIMULATION RESULTS

Cramér-Rao lower bounds indicate the best possible performance for this parameter estimation problem. Hence the performance of the proposing method can be evaluated by checking the gap between the solution and the theoretical minimum.

4.1. Cramér-Rao Lower Bounds

The noisy samples (8) can be re-expressed as:

$$\hat{y}_n = f(\mathbf{\Theta}, n) + \epsilon_n \tag{13}$$

where $\Theta = [x_1, x_2, \dots, x_K, t_1, t_2, \dots, t_K]^T$ is a vector of parameters to estimate, $n = 1, \dots, N$ and $f(\Theta, n) = \sum_{k=1}^{K} x_k \varphi(\tau_n - t_k)$.

Since any unbiased estimate of the unknown parameters Θ has a covariance matrix that is lower bounded by the inverse of the Fisher information matrix which is in this case

given by [11, 3]:

$$\operatorname{CRB}(\boldsymbol{\Theta}) = \sigma^2 \left(\sum_{n=1}^N \nabla f(\boldsymbol{\Theta}, n) \nabla f(\boldsymbol{\Theta}, n)^T \right)^{-1}.$$
 (14)

We are therefore able to evaluate numerically the CR bounds using the above equation.

4.2. Simulation Results

We first show in Figure 2 an example of the noiseless and noisy samples (PSNR = 15 dB, where PSNR is the peak signal-to-noise ratio defined as $10 \log_{10} \frac{|x_1|^2}{\sigma^2}$) that we obtain for N = 20 and for two different choices of a. We have then implemented the algorithm proposed in Section 3 and show the results in Figure 3 and 4.

In Figure 3, we use N = 50 samples to retrieve a single Dirac located at t = 0.5 sec. In this simulation, we have 100 realizations for each PSNR. The scatter plot of Figure 3



Fig. 2. Noisy and noiseless samples for two different hyperbolic secant kernels.

indicates that the estimation performance is satisfactory up to 15 dB for a proper a since the uncertainty on the location is quite close to the Cramér-Rao lower bound (see also Fig. 4 (d)).



Fig. 3. Scatter plot of the Dirac's location from N = 50 samples with a = 3. See also Figure 4 (d).

In Figures 4, we compare the standard deviation of our estimator with the CR bounds and this for different numbers of

 $^{^1 {\}rm This}$ can be achieved, for example, by performing the QR decomposition of ${\bf W}'_K.$

samples and for different values of a. Performance improves with the number of samples and this in line with the improvement in the corresponding CR bound. We notice that there is a gap between our estimation performance and the optimal performance when a is large. While the proposing algorithm cannot completely solve the problem of numerical issue arising from large a, the theory itself is tenable and works well for a proper a.



Fig. 4. Standard deviation of the proposed algorithm compared to Cramér-Rao bounds.

5. CONCLUSIONS

In this paper we have shown that it is possible to sample and perfectly reconstruct streams of Diracs using *non-uniformly* spaced samples. We have then presented a robust algorithm to reconstruct the Diracs when samples have been corrupted by noise. Preliminary results indicate that the algorithm performs well for medium to high SNRs. Future work will consider applying Cadzow iterative algorithm, as used in [3], to denoise the samples before retrieval of an FRI signal. We will also consider applying this sampling scheme to prospective applications.

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